

Low SNR analysis of the non-coherent MIMO channel under arbitrary channel and noise correlation structures

Marko Beko, João Xavier and Victor Barroso

Instituto de Sistemas e Robótica (ISR) – Instituto Superior Técnico

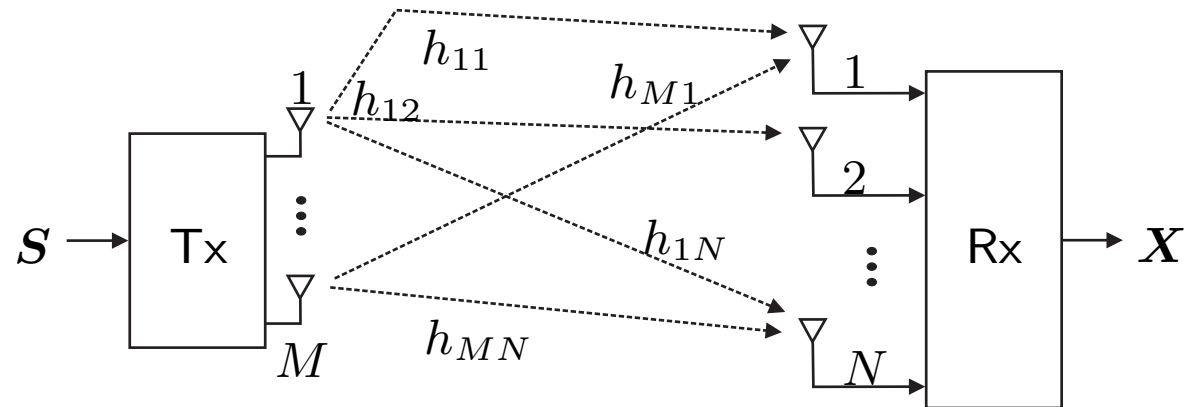
Av. Rovisco Pais, 1049–001

Lisboa, Portugal

{marko,jxavier,vab}@isr.ist.utl.pt

Problem Formulation

- ▷ Data model: $\mathbf{X} = \mathbf{S}\mathbf{H} + \mathbf{E}$



- ▷ $\mathbf{X}, \mathbf{E}: T \times N$, $\mathbf{S}: T \times M$, $\mathbf{H}: M \times N$
- ▷ Contribution: mutual information analysis for on-off and Gaussian signaling when $\mathbf{H} = \sqrt{\frac{\rho}{M}} \mathbf{K}^{\frac{1}{2}} \mathbf{H}_w$ and $\text{vec}(\mathbf{E}) \sim \mathcal{CN}(\mathbf{0}, \mathbf{\Upsilon})$ (colored noise)

Mutual information: on-off signaling

▷ The on-off signaling: for any $\epsilon > 1$,

$$\mathbf{S} = \mathbf{S}_{on} \rho^{-\frac{\epsilon}{2}} \text{ w.p. } \rho^\epsilon; \quad \mathbf{S} = \mathbf{0} \text{ w.p. } 1 - \rho^\epsilon$$

▷ At sufficiently low SNR

$$I(\mathbf{X}; \mathbf{S}) = \frac{\rho}{M} \text{tr} \left(\mathbf{\Upsilon}^{-1} (\mathbf{I}_N \otimes \mathbf{S}_{on}) \mathbf{K} \left(\mathbf{I}_N \otimes \mathbf{S}_{on}^H \right) \right) + o(\rho), \quad (1)$$

Mutual information: on-off signaling

▷ We maximize $I(\mathbf{X}; \mathbf{S})$ in (1) w.r.t \mathbf{S}_{on} and \mathbf{K}

$$\begin{aligned} & \max && \text{tr} \left(\mathbf{\Upsilon}^{-1} \bar{\mathbf{S}}_{on} \mathbf{K} \bar{\mathbf{S}}_{on}^H \right) && (2) \\ & \text{tr} \left(\mathbf{S}_{on} \mathbf{S}_{on}^H \right) \leq TM \\ & \mathbf{K} \in \mathcal{P}_{MN} \end{aligned}$$

where $\mathcal{P}_n = \{ \mathbf{Y} : n \times n \text{ matrix such that } \mathbf{Y} \succeq \mathbf{0}, \text{tr}(\mathbf{Y}) = n \}$
and $\bar{\mathbf{S}}_{on} = \mathbf{I}_N \otimes \mathbf{S}_{on}$.

▷ The maximum in (1) is attained by

$$\widehat{\mathbf{S}}_{on} = \text{ivector}(\widehat{\mathbf{s}}_{on}), \widehat{\mathbf{K}} = MN\widehat{\mathbf{u}}\widehat{\mathbf{u}}^H, \quad (3)$$

where

$$\begin{aligned} (\widehat{\mathbf{u}}, \widehat{\mathbf{s}}_{on}) = & \arg \max_{\substack{\|\mathbf{u}\| = 1, \\ \|\mathbf{s}_{on}\| \leq \sqrt{TM}}} \mathbf{s}_{on}^H \mathbf{M}_{on} \mathbf{s}_{on} \end{aligned} \quad (4)$$

with $\mathbf{M}_{on} = \mathbf{K}_{on}^H \left((\mathbf{u}\mathbf{u}^H)^T \otimes \mathbf{\Upsilon}^{-1} \right) \mathbf{K}_{on}$, $\mathbf{u} \in \mathbb{C}^{MN}$ and $\mathbf{s}_{on} = \text{vec}(\mathbf{S}_{on}) \in \mathbb{C}^{TM}$. $TMN^2 \times TM$ matrix \mathbf{K}_{on} is such that

$$\text{vec}(\mathbf{I}_N \otimes \mathbf{S}_{on}) = \mathbf{K}_{on} \text{vec}(\mathbf{S}_{on})$$

Mutual information: on-off signaling

- ▷ The optimization problem in (4) always admits a solution (maximization of a continuous function over a compact set)
- ▷ For the choice in (3), the maximal mutual information (p.c.u) is equal to

$$\frac{1}{T}I(\mathbf{X}; \mathbf{S}) = \rho N M \hat{\lambda} + o(\rho).$$

where

$$\hat{\lambda} = \underline{\hat{\mathbf{s}}}^H \mathbf{K}_{on}^H \left(\left(\hat{\mathbf{u}} \hat{\mathbf{u}}^H \right)^T \otimes \mathbf{\Upsilon}^{-1} \right) \mathbf{K}_{on} \underline{\hat{\mathbf{s}}} \quad (5)$$

and $\underline{\hat{\mathbf{s}}} = 1/\sqrt{TM} \hat{\mathbf{s}}_{on}$.

Mutual information: on-off signaling

Fixed eigenvectors of K

- ▷ In practice, by changing the antenna separation one can control the eigenvalues of K , but not their eigenvectors.
- ▷ In this case, the optimization problem defined in (2) becomes

$$\begin{aligned} (\hat{\mathbf{S}}_{on}, \hat{\mathbf{\Lambda}}) = & \arg \max \quad \text{tr} \left(\mathbf{\Upsilon}^{-1} \overline{\mathbf{S}}_{on} \mathbf{U} \mathbf{\Lambda} \mathbf{U}^H \overline{\mathbf{S}}_{on}^H \right) \quad (6) \\ & \|\mathbf{S}_{on}\| \leq \sqrt{TM} \\ & \mathbf{\Lambda} \in \mathcal{D}_{MN} \end{aligned}$$

where $\mathcal{D}_n = \{\mathbf{Y} : n \times n \text{ diagonal matrix such that } \mathbf{Y} \succeq \mathbf{0} \text{ and } \text{tr}(\mathbf{Y}) = n\}$, and $\mathbf{U} \mathbf{\Lambda} \mathbf{U}^H$ is the EVD of K .

Mutual information: on-off signaling

Fixed eigenvectors of K

▷ The maximum in (6) is attained by

$$\widehat{\mathbf{S}}_{on} = \sqrt{TM} \text{ivec}(\mathbf{s}_{\max}), \quad \widehat{\mathbf{K}} = MN \mathbf{u}_{i^*} \mathbf{u}_{i^*}^H, \quad (7)$$

where \mathbf{u}_i 's are the eigenvectors of \mathbf{K} , \mathbf{s}_{\max} is a unit-norm eigenvector associated to the $\lambda_{\max}(\mathbf{M}_{on}^{i^*})$ with

$$\mathbf{M}_{on}^{i^*} = \mathbf{K}_{on}^H \left((\mathbf{u}_{i^*} \mathbf{u}_{i^*}^H)^T \otimes \mathbf{\Upsilon}^{-1} \right) \mathbf{K}_{on} \quad (8)$$

and

$$i^* = \arg \max_{i = 1, 2, \dots, MN} \lambda_{\max}(\mathbf{M}_{on}^i). \quad (9)$$

Mutual information: on-off signaling

Fixed eigenvectors of K

- ▷ For the choice in (7), the maximal mutual information (p.c.u.) is given by

$$\frac{1}{T} I(\mathbf{X}; \mathbf{S}) = \rho N M \lambda_{\max} \left(\mathbf{M}_{on}^{i*} \right) + o(\rho).$$

Mutual information: on-off signaling

▷ Conclusions:

- From (3) and (7) we see that \mathbf{K} should be of rank one
- Correlated Rayleigh fading channel is beneficial from capacity viewpoint. Gain of order M with respect to uncorrelated Rayleigh fading channel
- On-off signaling attains the known channel capacity
- Correlation in noise is beneficial too, $\hat{\lambda} \geq 1$

Mutual information: Gaussian modulation

- ▷ On-off signaling is unpracticable due to large peakiness of the input signal
- ▷ Let $\mathbf{s} = \text{vec}(\mathbf{S}) \sim \mathcal{CN}(\mathbf{0}, \mathbf{P})$. At sufficiently low SNR

$$I(\mathbf{X}; \mathbf{S}) = \frac{\rho^2}{2M^2} \text{tr} \left(\mathbb{E}[\mathbf{Z}^2] - (\mathbb{E}[\mathbf{Z}])^2 \right) + o(\rho^2) \quad (10)$$

where $\mathbf{Z} = \mathbf{\Upsilon}^{-\frac{1}{2}} \overline{\mathbf{S}}_{on} \mathbf{K} \overline{\mathbf{S}}_{on}^H \mathbf{\Upsilon}^{-\frac{1}{2}}$

Mutual information: Gaussian modulation

▷ We maximize $I(\mathbf{X}; \mathbf{S})$ in (10) w.r.t \mathbf{P} and \mathbf{K}

$$\max_{\mathbf{P} \in \mathcal{H}_{TM}, \mathbf{K} \in \mathcal{P}_{MN}} \text{tr} \left(\mathbb{E}[\mathbf{Z}^2] - (\mathbb{E}[\mathbf{Z}])^2 \right) \quad (11)$$

where $\mathcal{H}_n = \{\mathbf{P} : n \times n \text{ matrices such that } \mathbf{P} = \mathbf{P}^H \succeq \mathbf{0} \text{ and } \text{tr}(\mathbf{P}) \leq n\}$.

Mutual information: Gaussian modulation

▷ The maximum in (11) is attained by

$$\hat{\mathbf{P}} = \hat{\mathbf{s}}_{on} \hat{\mathbf{s}}_{on}^H, \quad \hat{\mathbf{K}} = MN \hat{\mathbf{u}} \hat{\mathbf{u}}^H, \quad (12)$$

where

$$\begin{aligned} (\hat{\mathbf{u}}, \hat{\mathbf{s}}_{on}) = & \arg \max & \mathbf{s}_{on}^H \mathbf{M}_{on} \mathbf{s}_{on} & (13) \\ & \|\mathbf{u}\| = 1, & & \\ & \|\mathbf{s}_{on}\| \leq \sqrt{TM} & & \end{aligned}$$

Mutual information: Gaussian modulation

- ▷ For the choice in (12), the maximal mutual information (p.c.u) is equal to

$$\frac{1}{T} I(\mathbf{X}; \mathbf{S}) = \frac{\rho^2}{2} N^2 T M^2 \hat{\lambda}^2 + o(\rho^2)$$

where

$$\hat{\lambda} = \underline{\hat{\mathbf{s}}}^H \mathbf{K}_{on}^H \left(\left(\underline{\hat{\mathbf{u}}} \underline{\hat{\mathbf{u}}}^H \right)^T \otimes \mathbf{\Upsilon}^{-1} \right) \mathbf{K}_{on} \underline{\hat{\mathbf{s}}}$$

and $\underline{\hat{\mathbf{s}}} = 1/\sqrt{TM} \hat{\mathbf{s}}_{on}$.

Mutual information: Gaussian modulation

Fixed eigenvectors of K

- ▷ When the eigenvalues of K are fixed then the maximum of (11) is attained by

$$\hat{P} = TM s_{\max} s_{\max}^H, \quad \hat{K} = MN u_{i^*} u_{i^*}^H, \quad (14)$$

where u_i 's are the eigenvectors of K , s_{\max} is a unit-norm eigenvector associated to the $\lambda_{\max} \left(M_{on}^{i^*} \right)$ with

$$M_{on}^{i^*} = K_{on}^H \left((u_{i^*} u_{i^*}^H)^T \otimes \Upsilon^{-1} \right) K_{on} \quad (15)$$

and

$$i^* = \arg \max_{i = 1, 2, \dots, MN} \lambda_{\max} \left(M_{on}^i \right). \quad (16)$$

Mutual information: Gaussian modulation

Fixed eigenvectors of K

- ▷ For the choice in (14), the maximal mutual information (p.c.u.) is given by

$$\frac{1}{T} I(\mathbf{X}; \mathbf{S}) = \frac{\rho^2}{2} N^2 T M^2 \lambda_{\max}^2 \left(\mathbf{M}_{on}^{i*} \right) + o(\rho^2).$$

Mutual information: Gaussian modulation

▷ Conclusions:

- From (12) and (14) we see that \mathbf{K} should be of rank one
- Correlated Rayleigh fading channel is beneficial from capacity viewpoint. Gain of order M^2N with respect to uncorrelated Rayleigh fading channel
- Correlation in noise is beneficial too, $\hat{\lambda} \geq 1$

References

- [1] M. Beko, J. Xavier, and V. Barroso, "Capacity and error probability analysis of non-coherent MIMO systems in the low SNR regime," in *Proc. of IEEE International Conference on Acoustics, Speech, and Signal Processing (ICASSP)*, Honolulu, HI USA, 2007.
- [2] M. Beko, J. Xavier, and V. Barroso, "Further results on the capacity and error probability analysis of non-coherent MIMO systems in the low SNR regime," accepted for publication in *IEEE Transactions on Signal Processing*.
- [3] A. Lozano, A. Tulino, and S. Verdú, "Multiple-antenna capacity in the low-power regime," *IEEE Trans. Inform. Theory*, vol. 49, pp. 2527-2544, Oct. 2003.
- [4] I. Abou-Faycal, M. D. Trott, and S. Shamai, "The capacity of discrete-time memoryless Rayleigh fading channels," *IEEE Trans. Inf. Theory*, vol. 47, no. 4, pp. 1290-1301, May 2001.
- [5] S. Verdú, "Spectral efficiency in the wideband regime," *IEEE Trans. Inform. Theory*, vol. 48, pp. 1319-1343, June 2002.
- [6] V. Prelov and S. Verdú, "Second-order asymptotics of mutual information,"

IEEE Trans. Inform. Theory, vol. 50, pp. 1567-1580, Aug. 2004.

- [7] B. Hajek and V. Subramaniam, "Capacity and reliability function for small peak signal constraints," *IEEE Trans. Inform. Theory*, vol. 48, pp. 828-839, Apr. 2002.
- [8] C. Rao and B. Hassibi, "Analysis of multiple-antenna wireless links at low SNR," *IEEE Trans. Inf. Theory*, vol. 50, no. 9, pp. 2123-2130, Sep. 2004.
- [9] X. Wu and R. Srikant, "MIMO channels in the low SNR regime: communication rate, error exponent and signal peakiness," in *Proc. of IEEE Information Theory Workshop*, 2004.
- [10] W. Weichselberger, M. Herdin, H. Ozelik, E. Bonek, "A stochastic MIMO channel model with joint correlation of both link ends," *IEEE Trans. Wireless Commun.*, vol. 5, no. 1, pp. 90-100, Jan. 2006.
- [11] J. M. Borwein, A. S. Lewis, *Convex Analysis and Nonlinear Optimization: Theory and Examples.*, Springer-Verlag, 2000.
- [12] J. R. Magnus and H. Neudecker, *Matrix Differential Calculus with Applications in Statistics and Econometrics.*, Revised Edition, John Wiley & Sons, 1999.