

Codebook design for the non-coherent GLRT receiver and low SNR MIMO block fading channel

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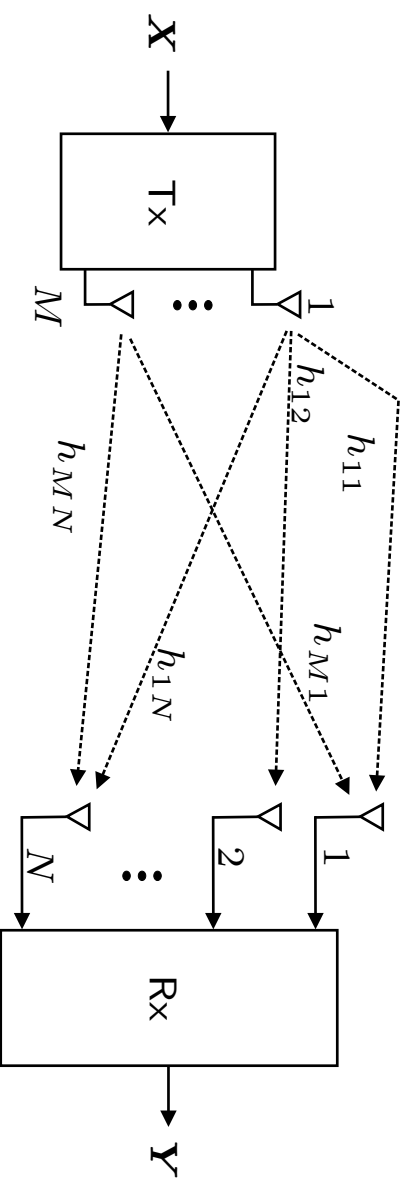
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Problem Formulation

▷ Data model: $\mathbf{Y} = \mathbf{X}\mathbf{H}^H + \mathbf{E}$



▷ \mathbf{Y}, \mathbf{E} : $T \times N$, \mathbf{x} : $T \times 1$, \mathbf{h} : $N \times 1$ for $M = 1$

▷ Codebook : $\mathcal{X} = \{\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_K\}$ is a point in the manifold

$$\mathcal{M} = \{(\mathbf{x}_1, \dots, \mathbf{x}_K) : \mathbf{x}_k^H \mathbf{x}_k = 1\}$$

▷ Contribution: PEP analysis and codebook design for $M = 1$ in low SNR regime when \mathbf{H} is deterministic and unknown

Problem Formulation

▷ GLRT receiver:

$$\begin{aligned}\hat{k} &= \operatorname{argmax}_{k=1,2,\dots,K} p(\mathbf{y}|\mathbf{x}_k, \hat{\mathbf{g}}_k) \\ &= \operatorname{argmin}_{k=1,2,\dots,K} \|\mathbf{y} - \widehat{\mathbf{X}}_k \hat{\mathbf{g}}_k\|\end{aligned}$$

$$\widehat{\mathbf{X}}_k = \mathbf{I}_N \otimes \mathbf{x}_k, \quad \hat{\mathbf{g}}_k = \widehat{\mathbf{X}}_k^H \mathbf{y} \quad (\text{ML channel estimate}), \quad \mathbf{y} = \operatorname{vec}(\mathbf{Y})$$

▷ PEP analysis: it can be shown that at sufficiently low SNR

$$P_{\mathbf{x}_i \rightarrow \mathbf{x}_j} = P \left(\sum_{n=1}^N (|a_n|^2 - |b_n|^2) > \|\mathbf{h}\|^2 \sin \alpha_{ij} \right) \quad (1)$$

where $a_n, b_n \stackrel{iid}{\sim} \mathcal{CN}(0, 1)$ and the angle α_{ij} is the acute angle between the codewords \mathbf{x}_i and \mathbf{x}_j

Problem Formulation

▷ In our work [5] the expression for the PEP in the high SNR regime and $M = 1$ is given by

$$P_{x_i \rightarrow x_j} = Q \left(\frac{1}{\sqrt{2}} \| \mathbf{h} \| \sin \alpha_{i,j} \right) \quad (2)$$

where $Q(x) = \int_x^{+\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}} dt$

- ▷ Equations (1)-(2) confirm that the codewords x_i and x_j should be constructed as separate as possible
- ▷ The problem of constructing good codes corresponds to the very well known packing problem in the complex projective space [4]

Problem Formulation

▷ Optimization problem: result (1) suggests the codebook design problem

$$\begin{aligned} \mathcal{X}^* &= \arg \max_{\mathcal{X} \in \mathcal{M}} \min \{L_{ij}(\mathcal{X}) : 1 \leq i \neq j \leq K\} \end{aligned} \quad (3)$$

where $L_{ij}(\mathcal{X}) = \mathbf{x}_i^H \mathbf{\Pi}_j^\perp \mathbf{x}_i = \sin^2 \alpha_{ij}$ with $\mathbf{\Pi}_j^\perp = \mathbf{I}_T - \underbrace{\mathbf{x}_j \mathbf{x}_j^H}_{\mathbf{\Pi}_j}$

▷ The problem in (3) is a high-dimensional, non-linear and non-smooth optimization problem

Example: for $K = 256$, $T = 8$, $M = 1$: $K(K - 1) = 65280$ $L_{ij}(\mathcal{X})$ functions and $2KT M = 4096$ real variables to optimize

Codebook Construction

- ▷ Two-phase methodology to tackle the optimization problem in (3) (see [5]-[6])
- ▷ Phase I: solves a convex semi-definite programming (SDP) relaxation
- ▷ Incremental approach: Let $\mathcal{X}_{k-1}^* = \{\mathbf{x}_1^*, \dots, \mathbf{x}_{k-1}^*\}$ be the codebook at the $k-1$ th stage. The new codeword is found by solving

$$\mathbf{x}_k^* = \arg \max_{\mathbf{x}_k^H \mathbf{x}_k = 1} \min_{1 \leq i \leq k-1} \{\lambda_{\min}(\mathbf{L}_{ik}), \lambda_{\min}(\mathbf{L}_{ki})\} \quad (4)$$

for $k = 2, \dots, K$

Codebook Construction - Phase I

▷ The optimization problem (4) is equivalent to (see [6])

$$(\hat{\mathbf{Y}}^*, \widetilde{\mathbf{X}}^*, t^*) = \arg \max_t \quad (5)$$

with the following constraints

$$\begin{bmatrix} \text{tr}(\mathbf{N}_i \mathbf{A}_1 \hat{\mathbf{Y}} \mathbf{B}_1) - t & \cdots & \text{tr}(\mathbf{N}_i \mathbf{A}_{MN} \hat{\mathbf{Y}} \mathbf{B}_1) \\ \vdots & & \vdots \\ \text{tr}(\mathbf{N}_i \mathbf{A}_1 \hat{\mathbf{Y}} \mathbf{B}_{MN}) & \cdots & \text{tr}(\mathbf{N}_i \mathbf{A}_{MN} \hat{\mathbf{Y}} \mathbf{B}_{MN}) - t \end{bmatrix} \succeq \mathbf{0},$$

$$\begin{bmatrix} M & \mathbf{Z}_i \\ \mathbf{Z}_i^H & \mathbf{P}_i - t \mathbf{I}_{MN} \end{bmatrix} \succeq \mathbf{0} \forall 1 \leq i \leq k-1, \quad \mathbf{K} \hat{\mathbf{Y}} \mathbf{K}^H = \widetilde{\mathbf{X}}, \text{tr}(\widetilde{\mathbf{X}}) = 1,$$

$$\mathbf{f} \hat{\mathbf{Y}} \mathbf{f}^H = 1, \hat{\mathbf{Y}} = \hat{\mathbf{Y}}^H, \hat{\mathbf{Y}} \succeq \mathbf{0}, \text{rank}(\hat{\mathbf{Y}}) = 1$$

$$\text{and } \widetilde{\mathbf{X}} = \mathbf{x}_k \mathbf{x}_k^H, b^2 = 1, \hat{\mathbf{Y}} = \mathbf{z} \mathbf{z}^H, \mathbf{z} = \left[\text{vec}^T(\widetilde{\mathbf{X}}_k) b \right]^T,$$

$$\widetilde{\mathbf{X}}_k = \mathbf{I}_N \otimes \mathbf{x}_k.$$

▷ The matrices M, \mathbf{Z}_i — linear in $\hat{\mathbf{Y}}$

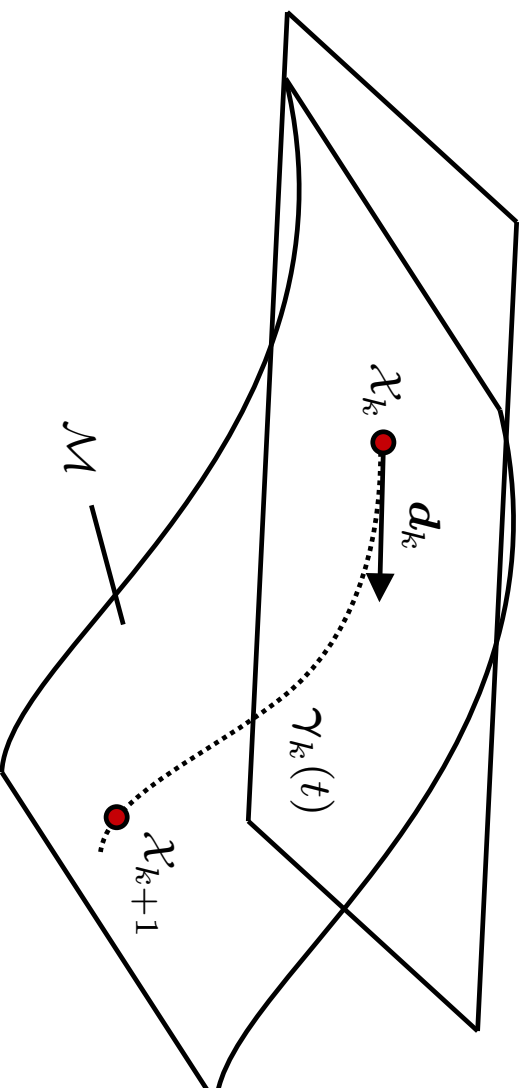
▷ The matrices $\mathbf{N}_i, \mathbf{P}_i, \mathbf{K}, \mathbf{f}, \mathbf{A}_i$ and \mathbf{B}_i — constants

Codebook Construction - Phase 1

- ▷ Design of the codewords: high-dimensional difficult nonlinear optimization problem (rank condition in (5))
- ▷ Relaxing the rank constraint leads to an SDP [7]
- ▷ The k^{th} codeword is extracted from the output variable $\widetilde{\mathbf{X}}$ with a technique similar to [8]
- ▷ Initialization \mathbf{x}_1^* : randomly generated

Codebook Construction - Phase 2

▷ Phase II: optimizes a non-smooth function on a manifold



Codebook Construction - Phase 2

- ▷ Iterative algorithm, called GDA (geodesic descent algorithm)
- ▷ Identify "active" pairs (i, j) that attain minimum in (3)
- ▷ Check if there is an ascent direction $d_k \in T_{\mathcal{X}_k} \mathcal{M}$ for all active (i, j) (consists of solving LP)
- ▷ When d_k is found, perform Armijo rule along geodesic $\gamma_k(t)$
- ▷ If no d_k is found, the algorithm stops

Computer Simulations

- Example: Constellations with equal priors

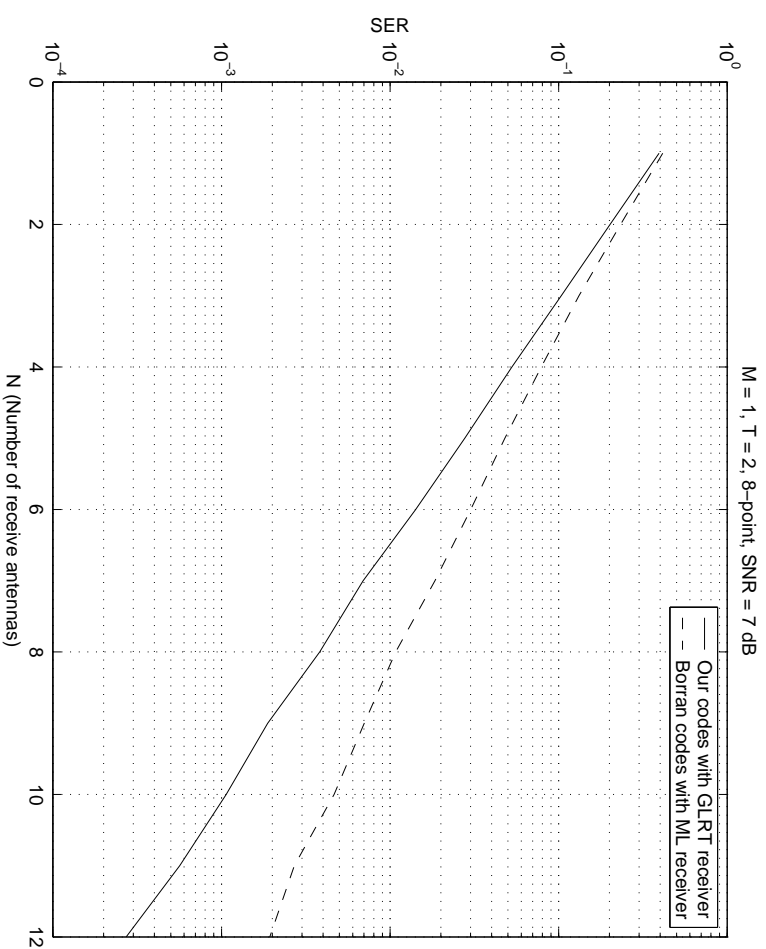


Figure 1: $M=1$, $T=2$, $K=8$, $\text{SNR} = 7$ dB. Solid curve:our codes with our GLRT receiver. Dashed curve: Borran codes designed for $\text{SNR} = 7$ dB with ML receiver [1].

□ Example: Constellations with equal priors

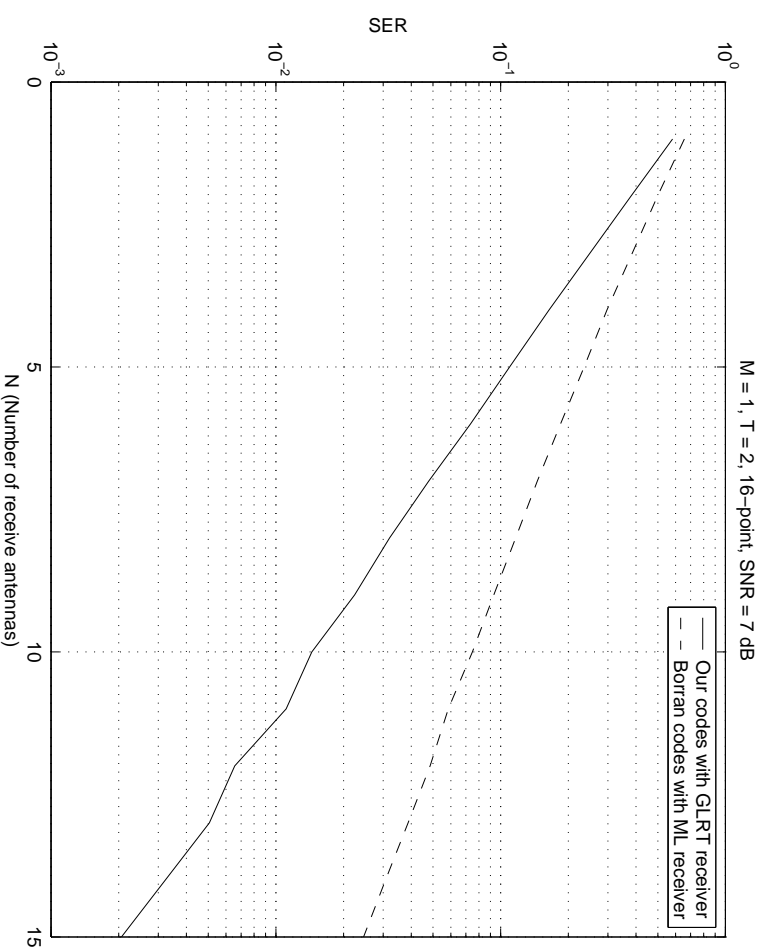


Figure 2: $M=1$, $T=2$, $K=16$, $\text{SNR} = 7$ dB. Solid curve:our codes with our GLRT receiver. Dashed curve: Borran codes designed for $\text{SNR} = 7$ dB with ML receiver [1].

□ Example: Constellations with unequal priors

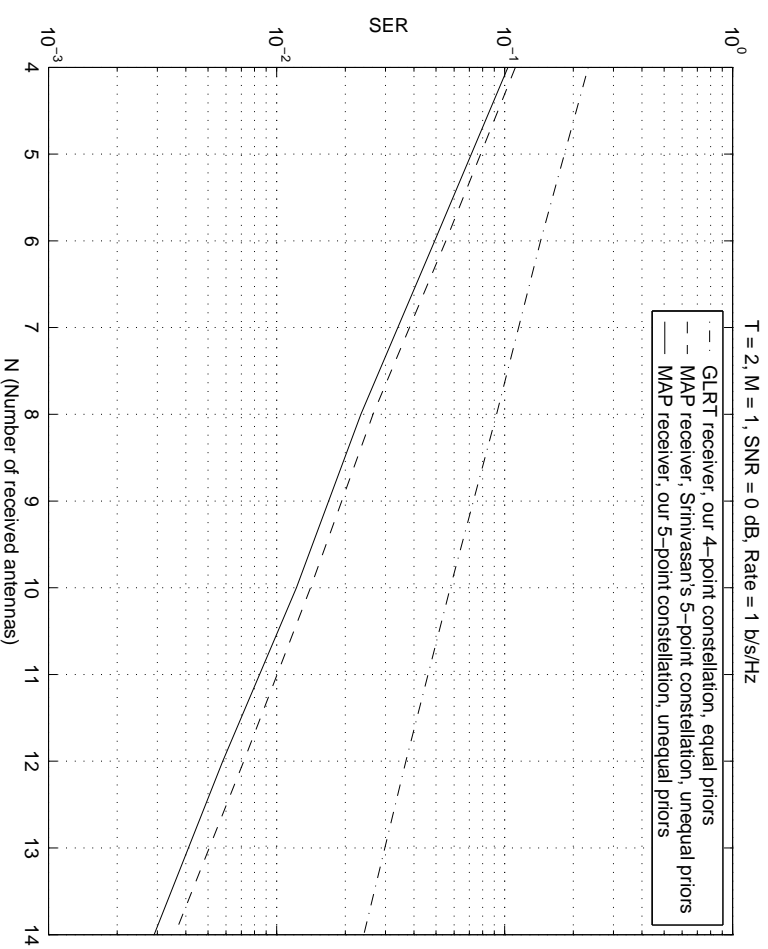


Figure 3: $T=2$, $M=1$, $\text{SNR} = 0 \text{ dB}$, $\text{Rate} = 1 \text{ b/s/Hz}$. Solid curve-our 5 point constellation with unequal priors, dashed curve-Srinivasan's 5 point constellation with equal priors [2], dash-dotted curve-our 4 point constellation with equal priors. Our and Srinivasan's 5 point constellations use *maximum a-posteriori* (MAP) receiver, our 4 point constellation uses GLRT receiver.

- Example: Constellations with equal priors and $M \geq 1$

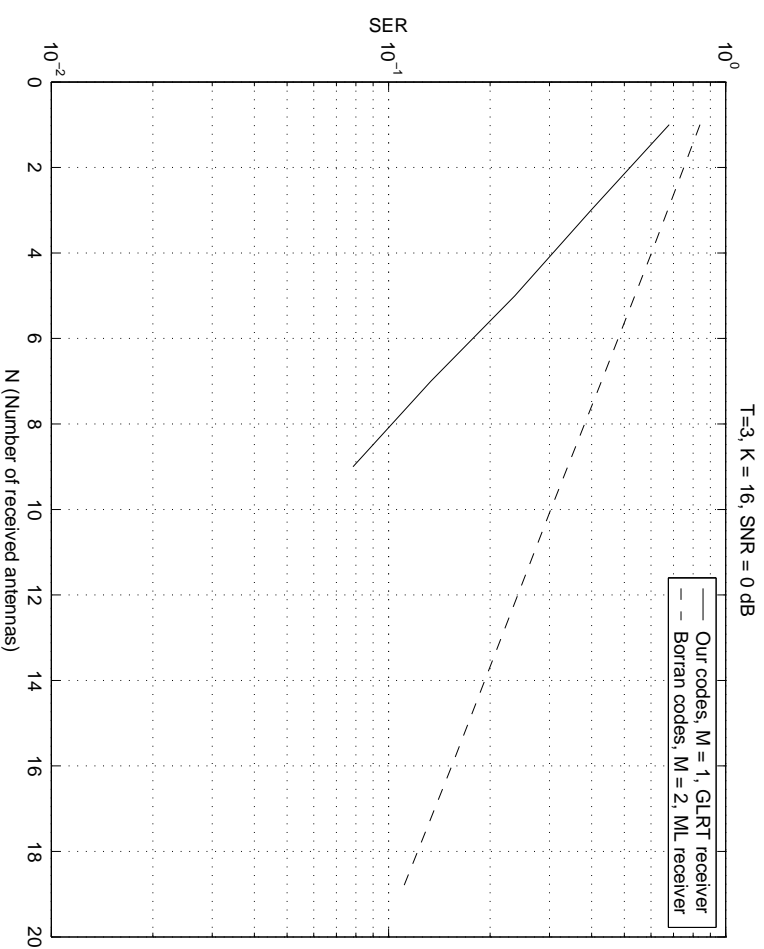


Figure 4: Solid curve-our codes for $K = 16$, $T = 3$, $M = 1$, dashed curve-Borran codes for $K = 16$, $T = 3$, $M = 2$.

□ Example: Constellations with equal priors and $M \geq 1$

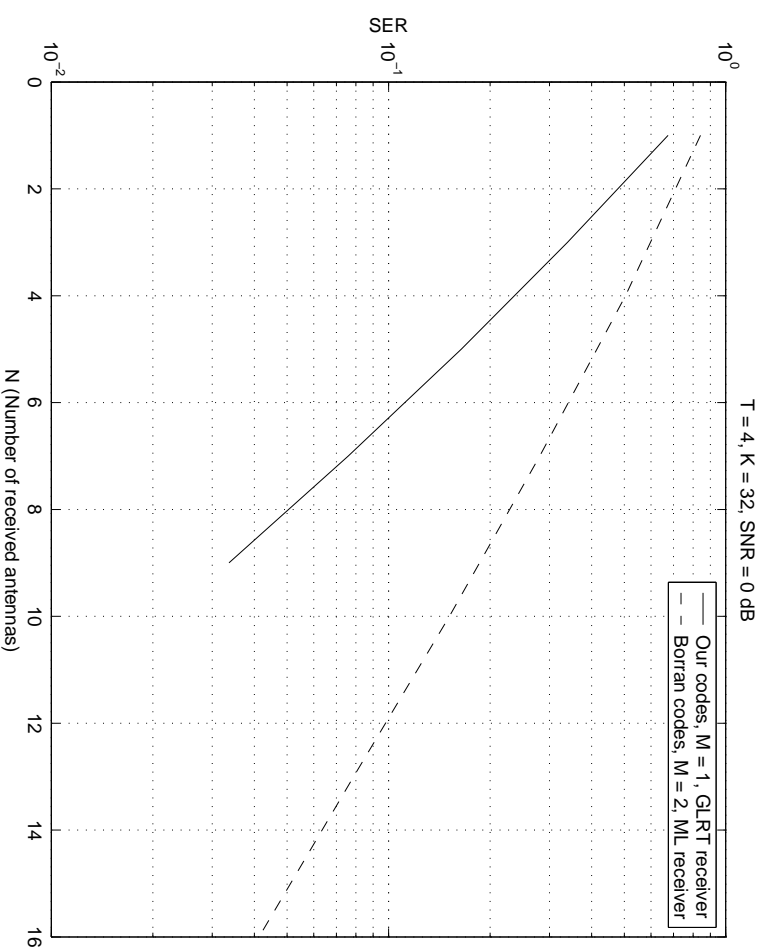


Figure 5: Solid curve-our codes for $K = 32$, $T = 4$, $M = 1$, dashed curve-Borran codes for $K = 32$, $T = 4$, $M = 2$.

□ Example: Constellations with equal priors and $M \geq 1$

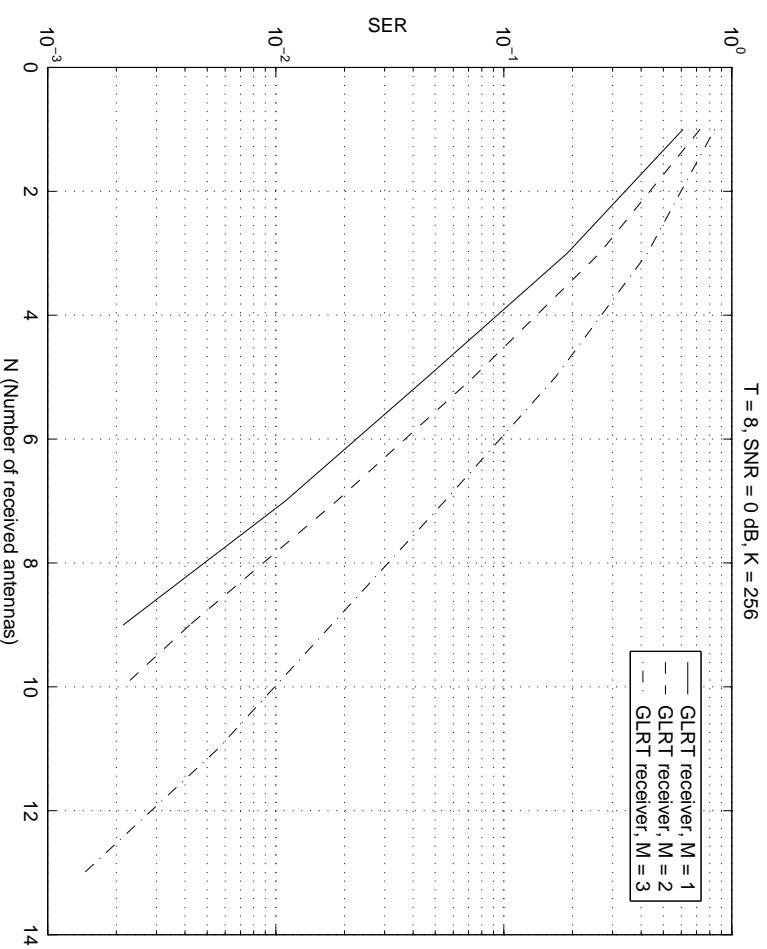


Figure 6: $T=8$, $K=256$, $\text{SNR} = 0 \text{ dB}$. Solid curve-our codes for $M = 1$, dashed curve-our codes for $M = 2$, dash-dotted curve-our codes for $M = 3$. All codes use GLRT receiver.

Conclusions

- ▷ PEP analysis and codebook design for $M = 1$ in low SNR regime when \mathbf{H} is deterministic and unknown
- ▷ Results
 - outperform significantly state-of-art known solutions which assume equal prior probabilities
 - also of interest for the constellations with unequal priors

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